

processing matrix of user $i \in \{1, 2, \dots, k, \dots, K\}$ in null space of previously coded matrix for $(i-1)$ th user. As each user faces different power constraint P_k , given by

$$P_k \geq \text{tr}(Q_k) \quad (5)$$

where Q_k is interference and noise covariance matrix, given by

$$Q_k = \sigma_{n_k}^2 I_{n_{R,k}} + E \{ [X_k X_k^H] | H \} \quad (6)$$

assuming interference is only contributed by precoded transmit processing matrix of previously added users, σ^2 being variance of random Gaussian variable z . σ^2 is 1 for $z \sim N(0,1)$ as proved mathematically in literature.

Using noise energy spectrum for k th user as in eq. (3), expanding eq. (6) further, we get

$$Q_i = \sigma_{n_k}^2 I_{n_{R,k}} + \sum_{j=1}^{i-1} (H_i W_j) (H_i W_j)^H \quad (7)$$

$$Q_i = \left(I_{n_{R,k}} + \frac{1}{\sigma_{n_k}^2} \sum_{j=1}^{i-1} (H_i W_j) (H_i W_j)^H \right) \quad (8)$$

The data rate for i th added user will be given by

$$R_i = \log_2 \left| I_{n_{T,k}} + \frac{1}{\sigma_{n_k}^2} H_i Q_i H_i^H \right| \quad (9)$$

eq (9) can be re-written in terms of SNR as

$$R_i = \log_2 \left| I_{n_{T,k}} + \frac{\gamma}{n_{R,k}} H_i H_i^H \right| \quad (10)$$

where γ is average SNR per user given by $P_k / \sigma_{n_k}^2$

The sum of data rate for all K users will be

$$R_{sum} = \log_2 \left| I_{N_T} + \sum_{k=1}^K \frac{\Gamma}{\sum_{k=1}^K n_{R,k}} H_k H_k^H \right| \quad (11)$$

where Γ is average SNR for all K users for total selected antennas given by $\sum_{k=1}^K n_{R,k}$

To maximize the data sum rate expression in eq. (11) while selecting antennas for k th user, keeping minimum threshold for SNR, requires maximization of

$$\sum_{i=1}^K H_k H_k^H = \|H_k\|_F^2 = \|H\|_F^2 \quad (12)$$

which is frobenius norm.

So individual antenna from pool of N_R , user antennas can be selected if the frobenius norm as per eq. (12) for the channel row is maximum for that antenna thus contributing to maximization of data sum rate viz. throughput.

E. Algorithm for combined user and antenna selection

Assuming set of users as $\mathbf{U} = \{1, 2, \dots, K\}$, and possible subsets of selected users at any time slot for which channel is static, is given by \mathbf{U}_l where $l = \{1, 2, \dots, \binom{K}{K_0}\}$, for K_0 number of users selected from K .

Number of antennas on each user and transmitter is same but denoted by N_R and N_T respectively for reason specified above. \mathcal{A}_j , represent the possible subset of antennas for k th user. $|\mathcal{A}_j| = N_R$, maximum number of antennas available on user, $j = \{1, 2, \dots, \binom{N_R}{n_{R,k}}\}$ where $n_{R,k}$ antennas are selected out of N_R antennas.

$$H = \{h_1, h_2, \dots, h_l, \dots, h_{n_{R,k}}\} \quad (13)$$

where in eq(13) $h_l \forall l$ is l th row of matrix given by

$$h_l = \{h_{l1}, h_{l2}, \dots, h_{ll}, \dots, h_{ln_{R,k}}\} \quad (14)$$

$$h_{ll} = \left\{ \begin{bmatrix} h_{l1} & \dots & h_{lN_{Tl}} \\ \vdots & \ddots & \vdots \\ h_{lN_{Rl}} & \dots & h_{lN_{Rl}N_{Tl}} \end{bmatrix} \in \mathcal{C}^{n_{Rl} \times n_{Tl}} \right\} \quad (15)$$

where h_{ll} is channel matrix for l th user

considering n_{Tl} beamforming transmit antennas selected for user l and n_{Rl} user antennas selected at user side. For selecting users and antennas, find $\|h_l\|_F^2 \forall l$ in \mathbf{U}_l .

The algorithm has following steps:

Step 1 : identify row h_l from eq. (11), giving best frobenius norm resulting in selection of l th user

Step 2: for l th user select n_{Rl} user antennas and n_{Tl} transmit antennas

- selecting n_{Rl} antennas at l th user
- from matrix in eq. (15) select column $u < v$ column iff $\|h_u\|^2 > \|h_v\|^2$, otherwise choose v th column so as to maximize the equation

$$\max_{2 \leq u, v \leq n_{Tl}} \left\{ \sum_{i=1}^{n_{Rl}} |h_{i,u}|^2 + |h_{i,v}|^2 \right\} \quad u \neq v$$

delete u th or v th column, repeat till we check all columns, and finally select n_{Rl} antennas

- repeat above step for rows to select n_{Tl} transmit antennas

$$\max_{2 \leq u, v \leq n_{R_l}} \left\{ \sum_{i=1}^{n_{R_l}} |h_{u,j}|^2 + |h_{v,j}|^2 \right\} u \neq v$$

delete u_{th} or v_{th} rows, repeat till we check all rows, and finally select n_{T_l} antennas

RESULTS AND DISCUSSION

Simulations have been carried to evaluate the performance of MU-MIMO in VANET. Simulations have been done with 500 different channel realizations considering linear precoding and singular value decomposition (SVD).

Table 1: Parameters for MU-MIMO VANET simulations

Parameters	Values
Frequency	5.9 GHz
Bandwidth	40 MHz/160 MHz
Modulation Scheme	64 QAM/256 QAM
Convolution Code	5/6
Channel Model	Nakagami -m fading model
Number of Antennas	8
Payload Size	100 bytes
Beacon Time Interval	100 ms
Signal to Noise Ratio	20 dB
Transmission Range	250-300 m
Area of Simulation	1000 x 1000 m ²
Number of Vehicles	10/20/40/60/80
Speed of Vehicles	20 -100 km/hr

The analytical results may have limited validity in practical scenario, so we have considered the vehicular movement scenario on urban and highway roads with vehicles ranging from 10 to 80 vehicles in 1000 x 1000 m² area with vehicle speed ranging from 20 km/hr to 100 km/hr considering the maximum permissible speed limit on urban roads and practical speeds on highway. Results have been obtained for safety messages with packet size of 100 bytes with beacon time interval of 100 ms, without much change in topology during message transmission, as the time taken for safety message dissemination are in order of microseconds analyzed in section IV. Transmission range of 250-300 m can simulate sparse traffic scenario with average of 1-2 vehicles in specified area. The antennas on all vehicular nodes are same, but for notational convenience, on transmitter it is denoted by

N_T and on users as N_R , with K users for VANET scenario. Maximum of 8 antennas have been considered on vehicle, mounted on roof-top in linear array. Performance has been evaluated in term of throughput, packet delivery ratio, delay and compared with results of SU-MIMO.

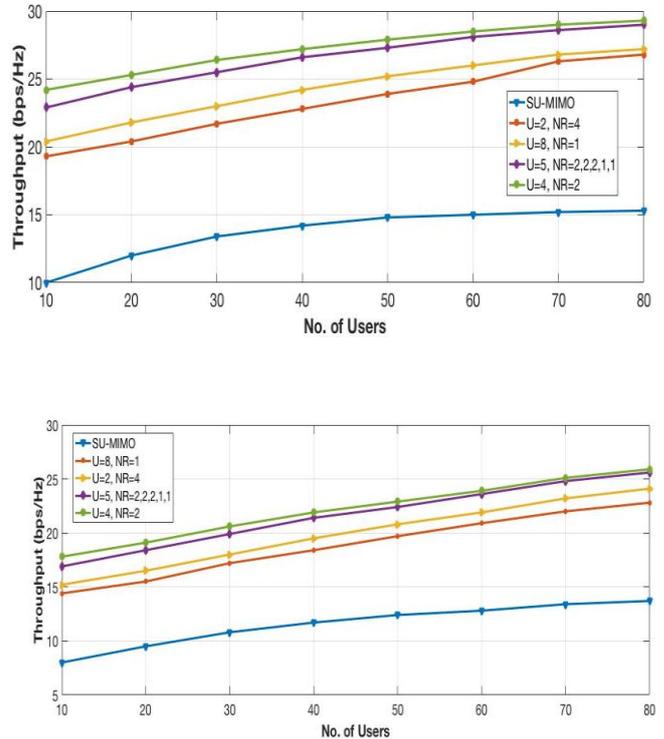


Figure 4: Throughput for SU-MIMO, and MU-MIMO for different number of user antennas and $N_T=8$ for a) urban road scenario b) highway scenario

Fig. 4 depicts the throughput for urban and highway scenario. SU in legend is for SU-MIMO, U2 for $K=2$, $n_{R_k} = 4$, $N_T=8$ in all cases. U8 for $K=8$, $n_{R_k} = 1$, U5 for $K=5$, $n_{R_k} = \{2,2,2,1,1\} \forall K = \{1,2,3,4,5\}$, U4 for $K=4$, $n_{R_k} = 2$ on each user. Performance of SU-MIMO is very low as compared to others. The scheme that make best use of multipath fading is benefitted most from MU-MIMO. Performance of U5 scheme is much closer to U4 for more number of users. Performance on highway scenario is low as multipath fading is less and LOS component is strong hence neutralizing the benefit of multipath fading for enhancing throughput.

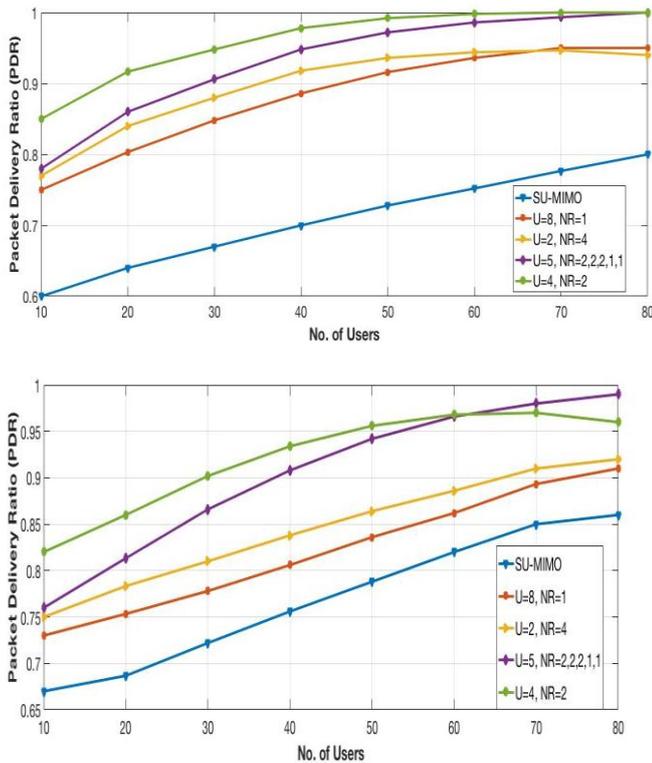


Figure 5: Packet Delivery Ratio for SU-MIMO, and MU-MIMO for different number of user antennas and $N_T=8$ for a) urban road scenario b) highway scenario

Fig. 5 shows the variation of packet delivery ratio (PDR) with number of antennas. PDR are the packets transmitted successfully versus the packets generated. PDR increases with increase in number of vehicles due to restricted mobility of vehicles resulting in successful packet transmission. User and antenna numbers are assumed to be same as in previous scenario. Transmission Scheme with 4 users having 2 antennas each outperforms others, as it exploits the benefit of multipath fading and spatial diversity owing to multiple users added in topology. The difference between SU-MIMO and MU-MIMO is wide in sparse vehicle scenario as the vehicles are distant located and connectivity becomes problem. Performance of all schemes in highway scenario is low in comparison to urban as a benefit of MIMO diversity due to buildings and obstacles creating multipaths is more prevalent on urban roads. Performance of MU-MIMO scheme with 4 users is better for sparse traffic from 5 users scheme than for dense traffic as the multipath diversity benefits are overpowered by congestion resulting in packet drops. PDR of 0.9 and above are required for reliable transmission in wireless networks but for VANETs it is difficult to achieve with SU-MIMO due to stringent network conditions as observed in fig. 5(a) and (b), but with MU-MIMO PDR above 0.9 is achievable.

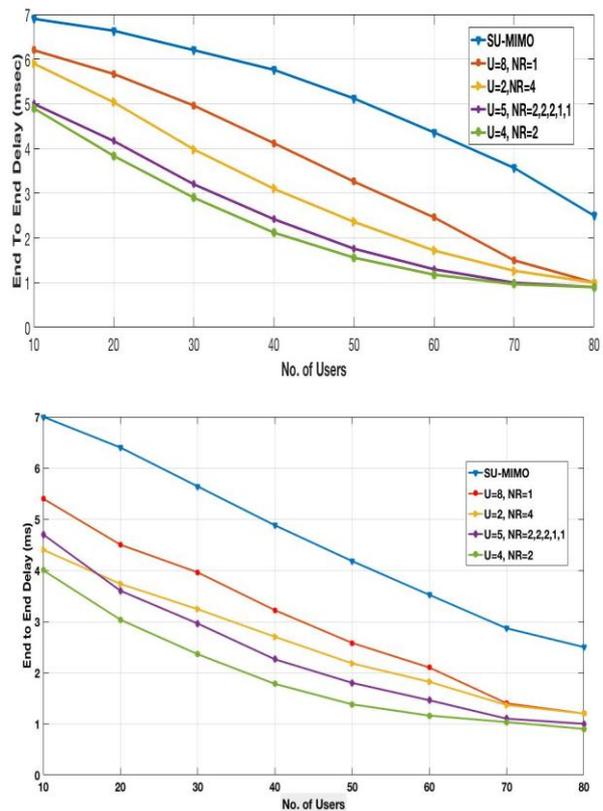


Figure 6: End to End for SU-MIMO, and MU-MIMO for different number of user antennas and $N_T=8$ for a) urban road scenario b) highway scenario

Fig. 6, compares the end to end delay in MU-MIMO VANET scenario. Delay is in order of milliseconds, and is least with 4 users having two selected antennas, giving maximum of 8 streams. For the obvious reason delay is more for sparse traffic as compared to dense, further as we reach to 80 vehicles on road it gives a breakeven and further increase in nodes may result in longer delay.

The results so obtained in terms of basic metrics are better for MU-MIMO, but involves complexity in terms of user and antennas selection. Hence it is worth analyzing the complexity of algorithm that is also one of the contributors in delay. For small number of antennas it may work but for the algorithm to work in wider perspective, scalability is important. Matrix calculations, frobenius norm calculations are major contributors in complexity.

In this algorithm vector calculations have to be done for finding frobenius norm to select users in step 1 for $U_l \times N_{R_l} \times N_{T_l}$ matrix given in eq. (15), vector calculations has complexity $O(U_l N_{R_l} N_{T_l})$. For step 2, vector calculations are required to find antennas for single user by selecting columns and then rows, adding complexity $O(N_{R_l} N_{T_l})$ and $O(n_{R_l} N_{T_l})$ respectively. For simplicity, selected users are represented by U and receive and transmit antennas, which are same on

vehicles is represented by N . Complexities of three vector calculations are $O(UN^2)$, $O(N^2)$, $O(N^2)$. Without loss of generality, considering large number of antennas and limited users $U \ll N$, the complexity can be given by $O(N^2)$ whereas the exhaustive search has complexity of the $O(N^3)$. Hence there is significant reduction in complexity as compared to exhaustive search hence contributes in delay reduction even for large number of antennas.

CONCLUSION

Realizing the underutilization of MIMO and visualizing the benefits of MU-MIMO in future applications, we have explored the possibility to implement MU-MIMO in VANET. The challenging task of user and antenna selection has been simplified by proposing joint user and antenna selection algorithm for VANET scenario. The results thus obtained in terms of throughput, PDR and end-to-end delay indicated MU-MIMO benefits in VANET. Throughput has doubled whereas PDR has shown significant increase as compared to SU-MIMO VANET, thus indicating the benefit of MU-MIMO for VANET. Further reduction in end-to-end delay to nearly half, indicate its benefit in VANET safety applications. Though the VANET realization has been done considering realistic scenario of urban and highway roads with sparse and dense traffic but due to limitations of simulator environment the practical results may vary but are still expected to be significantly better than SU-MIMO. The results obtained suggest the performance enhancement using MU-MIMO for applications other than safety in VANETs.

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